TPS3711 36-V Voltage Detector

1 Features

- Wide Supply Voltage Range: 1.8 V to 36 V
- Adjustable Threshold: Down to 400 mV
- Open-Drain Output for Undervoltage Detection
- Low Quiescent Current: 7 µA (Typical)
- High Threshold Accuracy:
  - 0.75% Over Temperature
  - 0.25% (Typical)
- Internal Hysteresis: 5.5 mV (Typical)
- Temperature Range: –40°C to +125°C
- Package: SOT-6

2 Applications

- Industrial Control Systems
- Embedded Computing Modules
- DSPs, Microcontrollers, and Microprocessors
- Notebook and Desktop Computers
- Portable- and Battery-Powered Products
- FPGA and ASIC Systems

3 Description

The TPS3711 wide-supply voltage comparator operates over a 1.8-V to 36-V range. The device has a precision comparator with an internal 400-mV reference and an open-drain output rated to 25 V for undervoltage detection. Set the monitored voltage with the use of external resistors.

OUT is driven low when the voltage at the SENSE pin drops below the negative threshold, and goes high when the voltage returns above the positive threshold. The comparator in the TPS3711 includes built-in hysteresis for noise rejection, thereby ensuring stable output operation without false triggering.

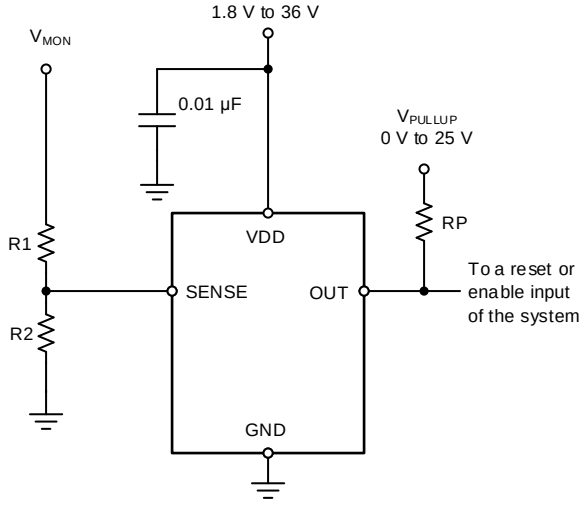
The TPS3711 is available in a SOT-6 package, and is specified over the junction temperature range of –40°C to +125°C.

Device Information (1)

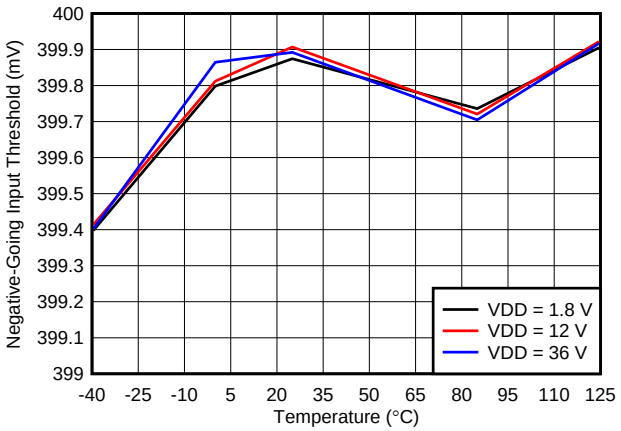
PART NUMBER	PACKAGE	BODY SIZE (NOM)
TPS3711	SOT (6)	2.90 mm × 1.60 mm

(1) For all available packages, see the package option addendum at the end of the datasheet.

Typical Application



Typical Error vs Junction Temperature



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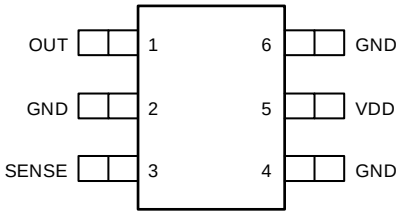
4 Revision History

Changes from Original (November 2015) to Revision A **Page**

- Changed input pin voltage maximum value from 1.7 V to 6.5 V ... 4
- Added tablenote ... 4

5 Pin Configuration and Functions

DDC Package
6-Pin SOT
Top View



Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
GND	2, 4, 6	—	Ground. Connect all three pins to ground.
OUT	1	O	Comparator open-drain output. This pin is driven low when the voltage at this comparator is less than V_{IT-}. The output goes high when the sense voltage rises above V_{IT+}.
SENSE	3	I	Comparator input. This pin is connected to the voltage to be monitored with the use of an external resistor divider. When the voltage at this pin drops below the threshold voltage V_{IT-}, OUT is driven low.
VDD	5	I	Supply-voltage input. Connect a 1.8-V to 36-V supply to VDD to power the device. It is good analog design practice to place a 0.1-µF ceramic capacitor close to this pin.

6 Specifications

6.1 Absolute Maximum Ratings (1)

over operating junction temperature range (unless otherwise noted)

		MIN	MAX	UNIT
Voltage (2)	V_{DD}	–0.3	40	V
	V_{OUT}	–0.3	28	
	V_{SENSE}	–0.3	7	
Current	Output pin current		40	mA
Temperature	Operating junction, T_J	–40	125	°C
	Storage, T_{stg}	-40	125	

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltages are with respect to network ground terminal.

6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	V
		Charged-device model (CDM), per JEDEC specification JESD22-C101 (2)	±500	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating junction temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_{DD}	Supply pin voltage	1.8		36	V
V_{SENSE}	Input pin voltage	0		6.5 (1)	V
V_{OUT}	Output pin voltage	0		25	V
V_{PULLUP}	Pullup voltage	0		25	V
I_{OUT}	Output pin current	0		10	mA
T_J	Junction temperature	–40	25	125	°C

(1) Operating V_{sense} at 1.7 V or higher and at 125°C continuously for 10 years or more would cause a degradation of accuracy spec to 1.5% maximum.

6.4 Thermal Information

THERMAL METRIC (1)		TPS3711 DDC (SOT) 6 PINS	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	201.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	47.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	51.2	°C/W
ψ_{JT}	Junction-to-top characterization parameter	0.7	°C/W
ψ_{JB}	Junction-to-board characterization parameter	50.8	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report (SPRA953).

6.5 Electrical Characteristics

Over the operating temperature range of T_J = –40°C to +125°C, 1.8 V ≤ V_{DD} < 36 V, and pullup resistor RP = 100 kΩ (unless otherwise noted). Typical values are at T_J = 25°C and V_{DD} = 12 V.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{(POR)}$	Power-on reset voltage(1)	V_{OL} ≤ 0.2 V			0.8	V
$V_{IT–}$	SENSE pin negative input threshold voltage	V_{DD} = 1.8 V to 36 V	397	400	403	mV
V_{IT+}	SENSE pin positive input threshold voltage	V_{DD} = 1.8 V to 36 V	400	405.5	413	mV
V_{HYS}	SENSE pin hysteresis voltage ($HYS = V_{IT+} – V_{IT–}$)		2	5.5	12	mV
V_{OL}	Low-level output voltage	V_{DD} = 1.8 V, I_{OUT} = 3 mA		130	250	mV
		V_{DD} = 5 V, I_{OUT} = 5 mA		150	250	
I_{IN}	Input current (at SENSE pin)	V_{DD} = 1.8 V and 36 V, V_{SENSE} = 6.5 V	–25	+1	+25	nA
		V_{DD} = 1.8 V and 36 V, V_{SENSE} = 0.1 V	–15	+1	+15	
$I_{D(leak)}$	Open-drain leakage current	V_{DD} = 1.8 V and 36 V, V_{OUT} = 25 V		10	300	nA
I_{DD}	Supply current	V_{DD} = 1.8 V – 36 V		8	11	µA
UVLO	Undervoltage lockout(2)	V_{DD} falling	1.3	1.5	1.7	V

(1) The lowest supply voltage (V_{DD}) at which output is active; $t_{r(VDD)}$ > 15 µs/V. If less than $V_{(POR)}$, the output is undetermined.
(2) When V_{DD} falls below UVLO, OUT is driven low. The output cannot be determined if less than $V_{(POR)}$.

6.6 Timing Requirements

PARAMETER		TEST CONDITION	MIN	TYP	MAX	UNIT
$t_{pd(HL)}$	High-to-low propagation delay(1)	V_{DD} = 24 V, ±10-mV input overdrive, R_L = 100 kΩ, V_{OH} = 0.9 × V_{DD}, V_{OL} = 250 mV		9.9		µs
$t_{pd(LH)}$	Low-to-high propagation delay(1)	V_{DD} = 24 V, ±10-mV input overdrive, R_L = 100 kΩ, V_{OH} = 0.9 × V_{DD}, V_{OL} = 250 mV		28.1		µs
$t_{d(start)}$(2)	Startup delay	V_{DD} = 5 V		155		µs
t_r	Output rise time	V_{DD} = 12 V, 10-mV input overdrive, R_L = 100 kΩ, C_L = 10 pF, V_O = (0.1 to 0.9) × V_{DD}		2.7		µs
t_f	Output fall time	V_{DD} = 12 V, 10-mV input overdrive, R_L = 100 kΩ, C_L = 10 pF, V_O = (0.9 to 0.1) × V_{DD}		0.12		µs

(1) High-to-low and low-to-high refers to the transition at the input pin (SENSE).
(2) During power on, V_{DD} must exceed 1.8 V for at least 150 µs (typ) before the output state reflects the input condition.

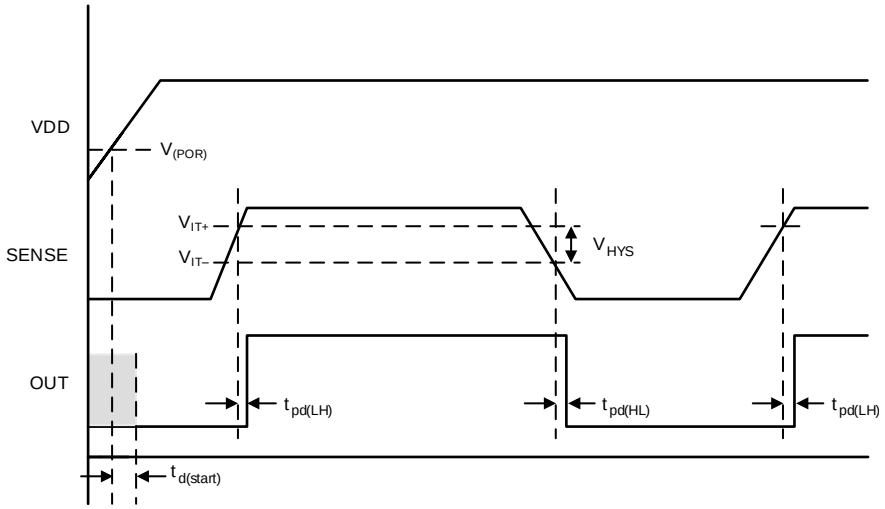


Figure 1. Timing Diagram

6.7 Typical Characteristics

at T_J = 25°C and V_{DD} = 12 V (unless otherwise noted)

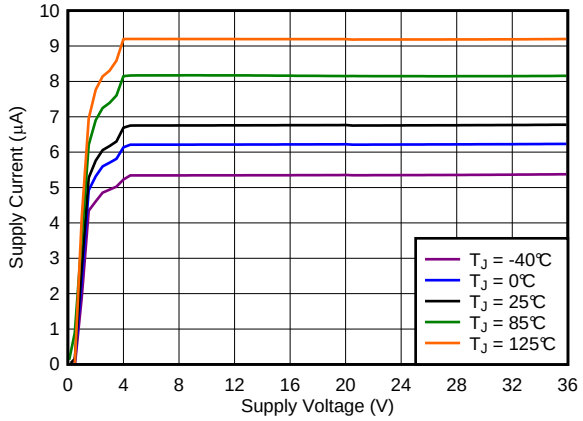
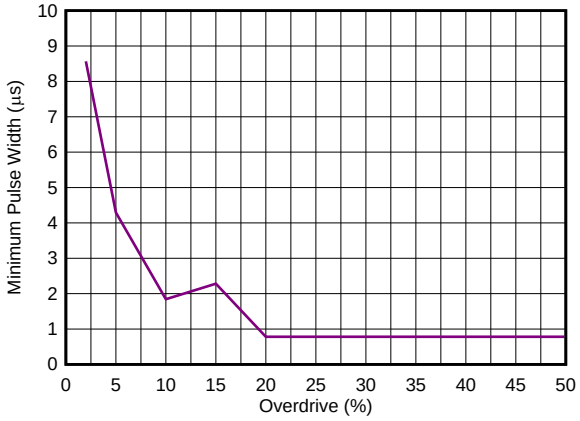


Figure 2. Supply Current vs Supply Voltage



V_{DD} = 24 V, minimum pulse duration required to trigger output high-to-low transition, SENSE = negative spike below V_{IT-}

Figure 3. Minimum Pulse Duration vs Threshold Overdrive Voltage

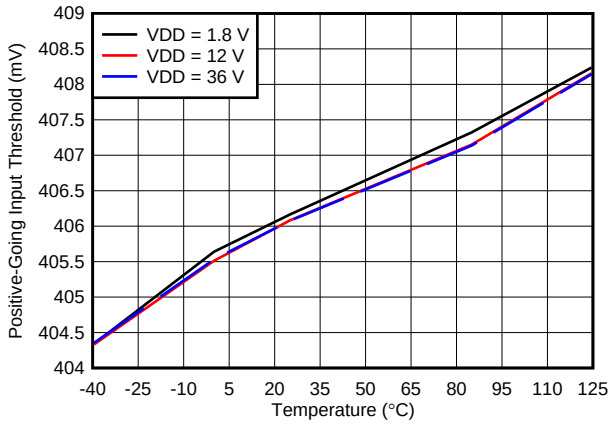


Figure 4. SENSE Positive Input Threshold Voltage (V_{IT+}) vs Temperature

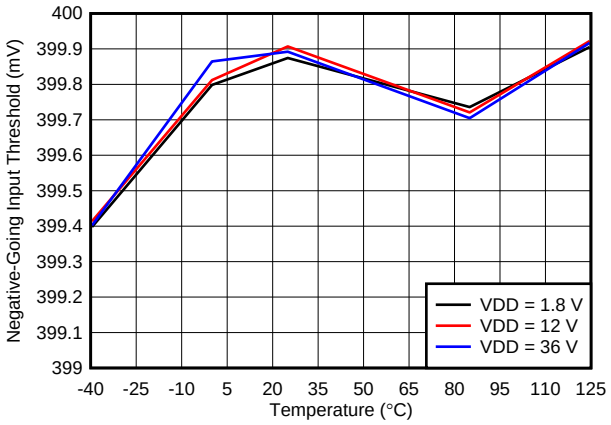
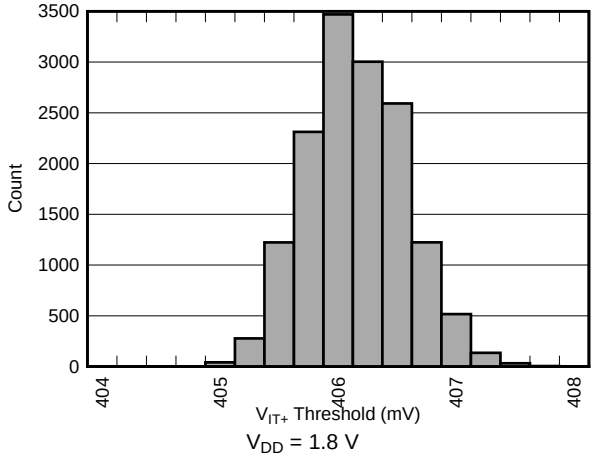
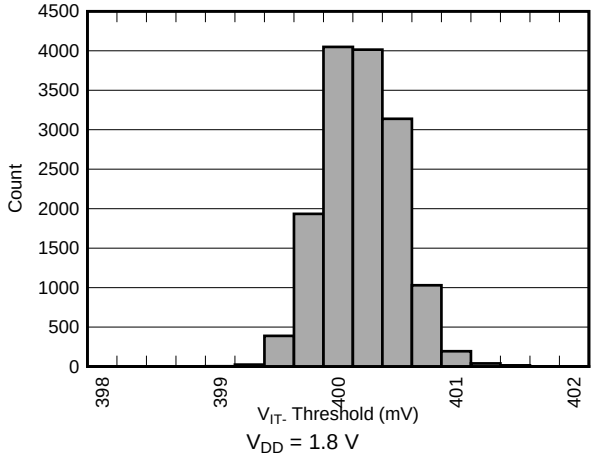


Figure 5. SENSE Negative Input Threshold Voltage (V_{IT-}) vs Temperature



V_{DD} = 1.8 V

Figure 6. SENSE Positive Input Threshold Voltage (V_{IT+}) Distribution

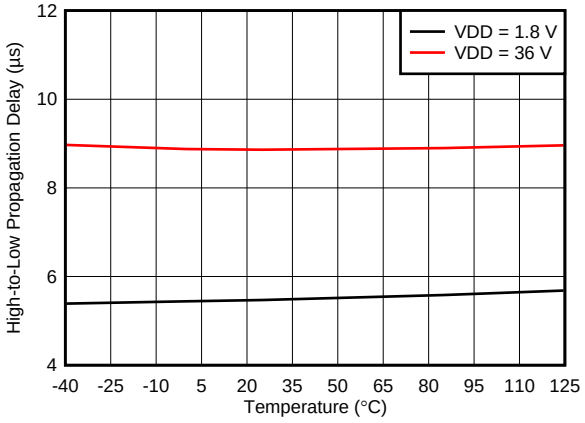


V_{DD} = 1.8 V

Figure 7. SENSE Negative Input Threshold Voltage (V_{IT-}) Distribution

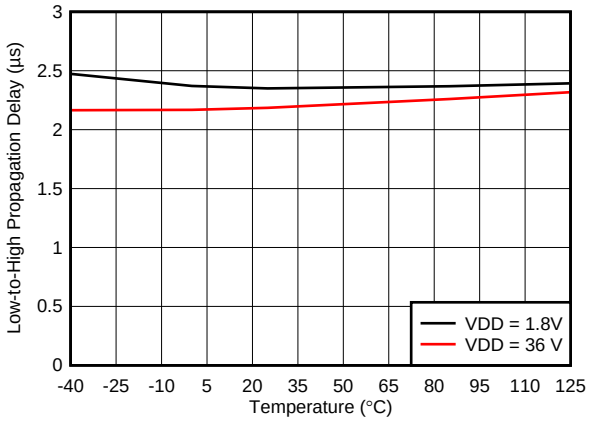
Typical Characteristics (continued)

at T_J = 25°C and V_{DD} = 12 V (unless otherwise noted)



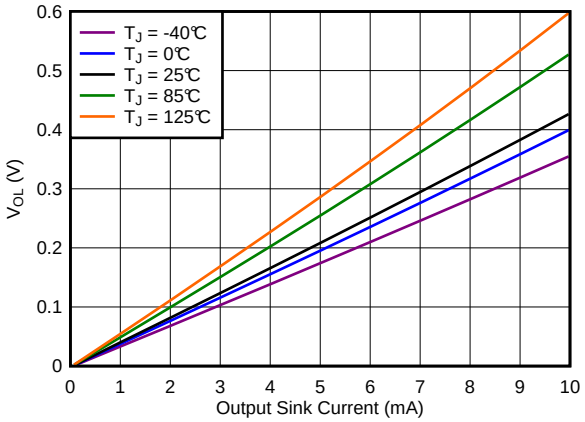
Input step ±200 mV

Figure 8. Propagation Delay vs Temperature (High-to-Low Transition at SENSE)



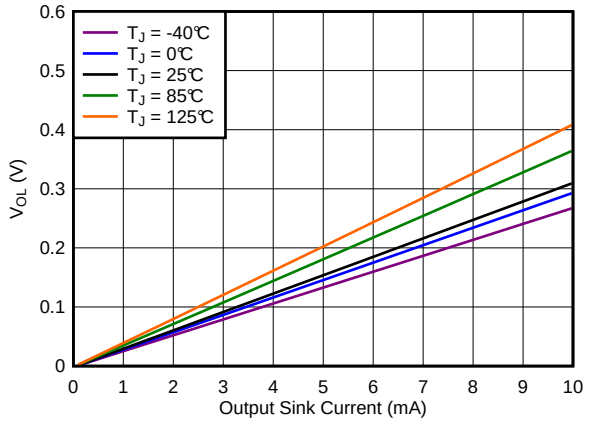
Input step ±200 mV

Figure 9. Propagation Delay vs Temperature (Low-to-High Transition at SENSE)



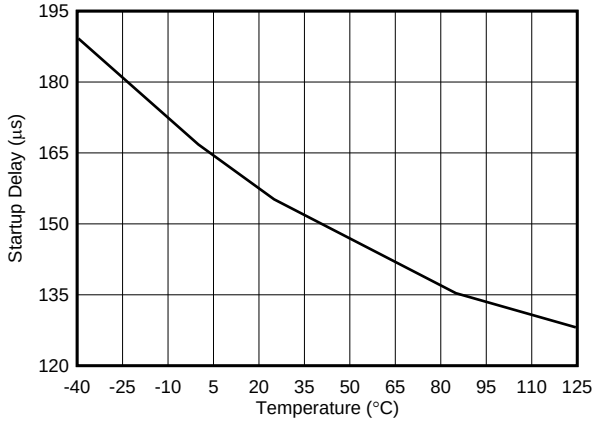
V_{DD} = 1.8 V

Figure 10. Output Voltage Low vs Output Sink Current



V_{DD} = 12 V

Figure 11. Output Voltage Low vs Output Sink Current



V_{DD} = 5 V

Figure 12. Startup Delay vs Temperature

7 Detailed Description

7.1 Overview

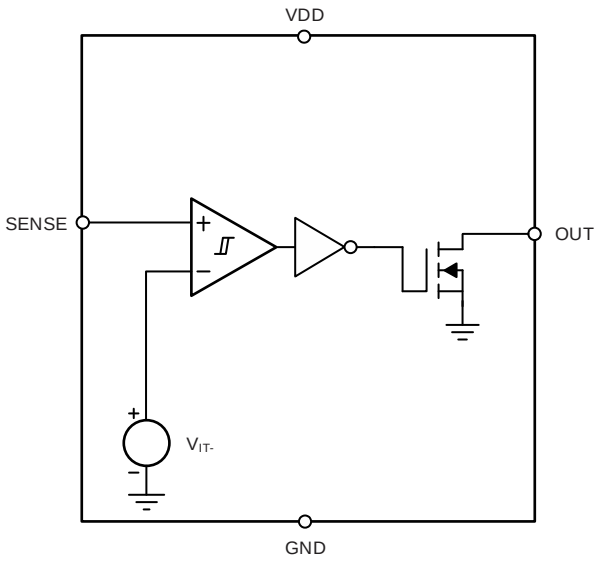
The TPS3711 combines a comparator and a precision reference for undervoltage detection. The TPS3711 features a wide supply voltage range (1.8 V to 36 V) and a high-accuracy threshold voltage of 400 mV (0.75% over temperature) with built-in hysteresis. The output is rated to 25 V and can sink up to 10 mA.

Set the input pin (SENSE) to monitor any voltage above 0.4 V by using an external resistor divider network. SENSE has very low input leakage current, allowing the use of a large resistor divider without sacrificing system accuracy. The relationship between the input and the output is shown in Table 1. Broad voltage thresholds are supported that enable the device to be used in a wide array of applications.

Table 1. Truth Table

CONDITION	OUTPUT	STATUS
SENSE > V_{IT+}	OUT high	Output high impedance
SENSE < V_{IT-}	OUT low	Output asserted

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 Input Pin (SENSE)

The TPS3711 combines a comparator with a precision reference voltage. The comparator has one external input and one internal input connected to the internal reference. The falling threshold on SENSE is designed and trimmed to be equal to the reference voltage (400 mV). This configuration optimizes the device accuracy. The comparator also has built-in hysteresis that proves immunity to noise and ensures stable operation.

The comparator input swings from ground to 6.5 V (7.0 V absolute maximum), regardless of the device supply voltage used. Although not required in most cases, it is good analog design practice to place a 1-nF to 10-nF bypass capacitor at the comparator input for noisy applications in order to reduce sensitivity to transient voltage changes on the monitored signal.

For the comparator, the output (OUT) is driven to logic low when the input SENSE voltage drops below V_{IT-}. When the voltage exceeds V_{IT+}, OUT goes to a high-impedance state; see Figure 1.

7.3.2 Output Pin (OUT)

In a typical TPS3711 application, the output is connected to a reset or enable input of the processor [such as a digital signal processor (DSP), application-specific integrated circuit (ASIC), or other processor type] or the output is connected to the enable input of a voltage regulator [such as a dc-dc converter or low-dropout regulator (LDO)].

The TPS3711 provides an open-drain output (OUT); use a pullup resistor to hold the line high when the output goes to a high-impedance state. Connect this pullup resistor to a voltage rail that meets the logic requirements of the downstream device. The TPS3711 output can be pulled up to 25 V, independent of the device supply voltage. To ensure the proper voltage level, give some consideration when choosing the pullup resistor value. The pullup resistor value is determined by V_{OL}, output capacitive loading, and the open-drain leakage current ($I_{D(leak)}$). These values are specified in the *Electrical Characteristics* table.

Table 1 and the *Input Pin (SENSE)* section describe how the output is asserted or high impedance. See Figure 1 for a timing diagram that describes the relationship between threshold voltage and the respective output.

7.4 Device Functional Modes

7.4.1 Normal Operation (V_{DD} > UVLO)

When the voltage on VDD is greater than 1.8 V for at least 155 µs, the OUT signal corresponds to the voltage on SENSE, as listed in Table 1.

7.4.2 Undervoltage Lockout ($V_{(POR)} < V_{DD} <$ UVLO)

When the voltage on VDD is less than the device UVLO voltage, and greater than the power-on reset voltage, $V_{(POR)}$, the OUT signal is asserted regardless of the voltage on SENSE.

7.4.3 Power On Reset ($V_{DD} < V_{(POR)}$)

When the voltage on VDD is lower than the required voltage to internally pull the asserted output to GND ($V_{(POR)}$), OUT is in a high-impedance state.

8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The TPS3711 is used as a precision voltage supervisor in several different configurations. The monitored voltage (V_{MON}), VDD voltage, and output pullup voltage can be independent voltages or connected in any configuration. The following sections show the connection configurations and the voltage limitations for each configuration.

8.1.1 Input and Output Configurations

Figure 13 to Figure 14 show examples of the various input and output configurations.

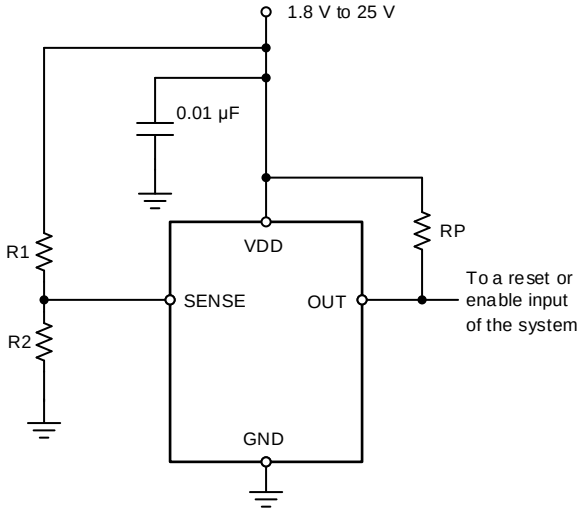
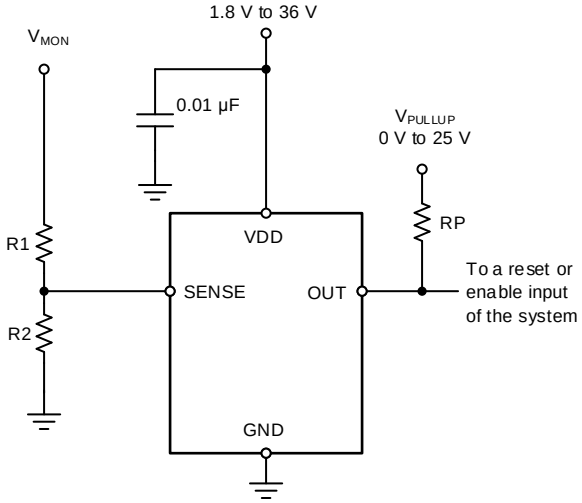


Figure 13. Monitoring the Same Voltage as V_{DD}



NOTE: The input can monitor a voltage higher than V_{DD} (max) with the use of an external resistor divider network.

Figure 14. Monitoring a Voltage Other than V_{DD}

Application Information (continued)

8.1.2 Immunity to Input Pin Voltage Transients

The TPS3711 is immune to short voltage transient spikes on the input pin. Sensitivity to transients depends on both transient duration and amplitude; see Figure 3, *Minimum Pulse Duration vs Threshold Overdrive Voltage*.

8.2 Typical Application

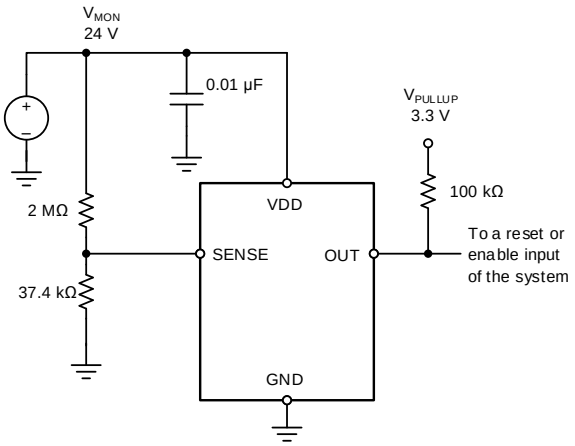


Figure 15. 24-V, 10% Comparator

8.2.1 Design Requirements

Table 2. Design Parameters

PARAMETER	DESIGN REQUIREMENT	DESIGN RESULT
Monitored voltage	24-V nominal, falling ($V_{MON(UV)}$) threshold 10% nominal (21.6 V)	$V_{MON(UV)}$ = 21.8 V ±2.7%
Output logic voltage	3.3-V CMOS	3.3-V CMOS
Maximum current consumption	30 µA	24 µA

8.2.2 Detailed Design Procedure

8.2.2.1 Resistor Divider Selection

The resistor divider values and target threshold voltage can be calculated by using Equation 1 to determine $V_{MON(UV)}$.

$$V_{MON(UV)} = \left(1 + \frac{R1}{R2}\right) \times V_{IT-} \quad (1)$$

where

- R1 and R2 are the resistor values for the resistor divider on the SENSE pin
- $V_{MON(UV)}$ is the target voltage at which an undervoltage condition is detected

Choose an R_{TOTAL} (= R1 + R2) so that the current through the divider is approximately 100 times higher than the input current at the SENSE pin. Use resistors with high values to minimize current consumption (as a result of low input bias current) without adding significant error to the resistive divider. For details on sizing input resistors, refer to application report SLVA450, *Optimizing Resistor Dividers at a Comparator Input*, available for download from www.ti.com.

8.2.2.2 Pullup Resistor Selection

To ensure the proper logic-high voltage level (V_{HI}), select a pullup resistor value where the pullup voltage divided by the pullup resistor value does not exceed the sink-current capability of the device. Confirm this voltage level by verifying that the pullup voltage minus the open-drain leakage current ($I_{D(leak)}$) multiplied by the resistor is greater than the desired V_{HI}. These values are specified in the *Electrical Characteristics* .

Use Equation 2 to calculate the value of the pullup resistor.

$$\frac{V_{HI} - V_{pullup}}{I_{D(leak)}} \leq RP \leq \frac{V_{pullup}}{I_{OUT}} \quad (2)$$

8.2.2.3 Input Supply Capacitor

Although an input capacitor is not required for stability, for good analog design practice, connect a 0.1-µF low equivalent series resistance (ESR) capacitor across the VDD and GND pins. A higher-value capacitor may be necessary if large, fast rise-time load transients are anticipated, or if the device is not located close to the power source.

8.2.3 Application Curves

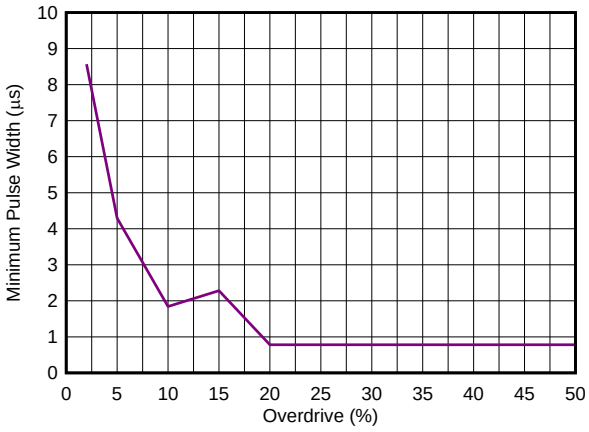


Figure 16. 24-V Window Monitor Output Response

9 Power Supply Recommendations

The TPS3711 has a 40-V absolute maximum rating on the VDD pin, with a recommended maximum operating condition of 36 V. If the voltage supply that provides power to VDD is susceptible to any large voltage transient that may exceed 40 V, or if the supply exhibits high voltage slew rates greater than 1 V/µs, then place an RC filter between the supply and VDD to filter any high-frequency transient surges on the VDD pin. In these cases, a 100-Ω resistor and 0.01-µF capacitor are required, as shown in Figure 17.

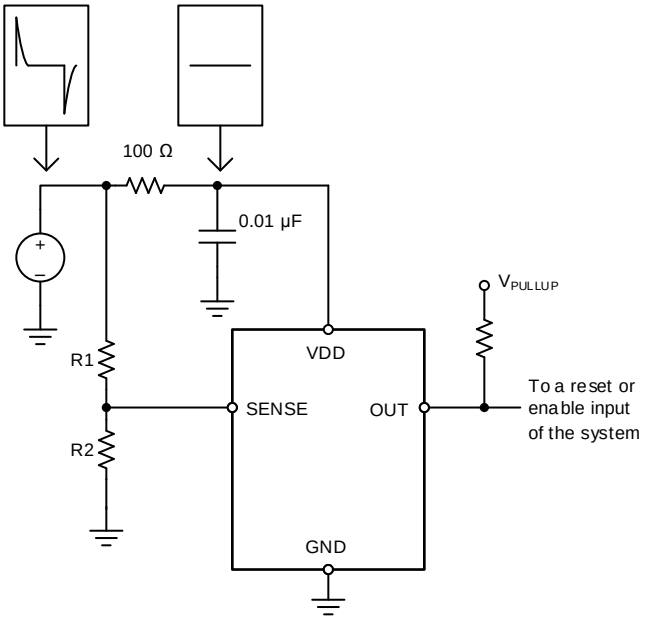


Figure 17. Using an RC Filter to Remove High-Frequency Disturbances on VDD

10 Layout

10.1 Layout Guidelines

- Place R_1 and R_2 close to the device to minimize noise coupling into the SENSE node.
- Place the VDD decoupling capacitor close to the device.
- Avoid using long traces for the VDD supply node. The VDD capacitor (C_{VDD}), along with parasitic inductance from the supply to the capacitor, might form an LC tank and create ringing with peak voltages above the maximum VDD voltage. If long traces are unavoidable, see Figure 17 for an example of filtering VDD.

10.2 Layout Example

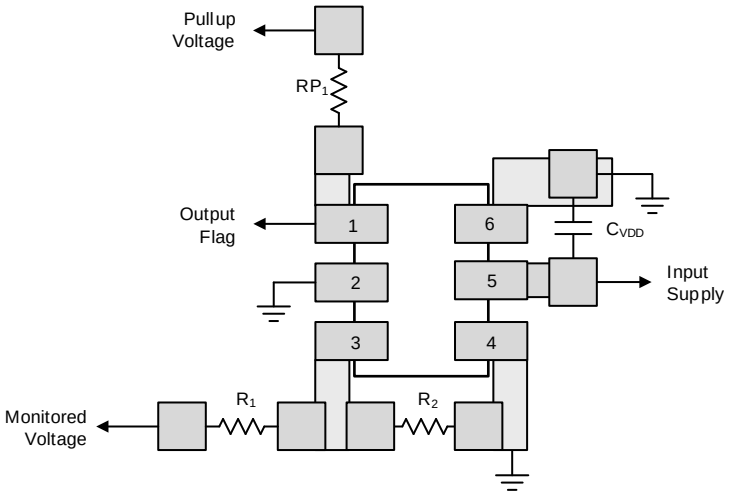


Figure 18. Recommended Layout

11 Device and Documentation Support

11.1 Documentation Support

11.1.1 Related Documentation

For related documentation, see the following application report, available through the TI website at www.ti.com:

- *Optimizing Resistor Dividers at a Comparator Input*, SLVA450

11.2 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community ***TI's Engineer-to-Engineer (E2E) Community.*** Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support ***TI's Design Support*** Quickly find helpful E2E forums along with design support tools and contact information for technical support.

11.3 Trademarks

E2E is a trademark of Texas Instruments.
All other trademarks are the property of their respective owners.

11.4 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

11.5 Glossary

SLYZ022 — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS3711DDCR	ACTIVE	SOT-23-THIN	DDC	6	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	11BO	Samples
TPS3711DDCT	ACTIVE	SOT-23-THIN	DDC	6	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	11BO	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

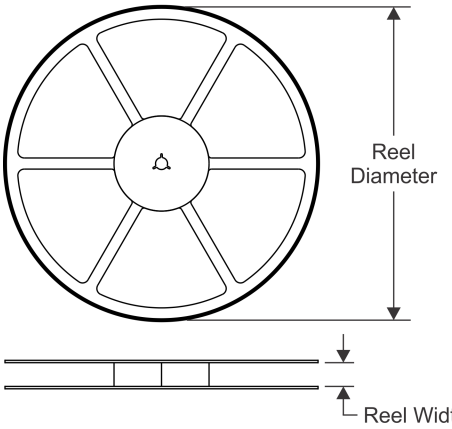
(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

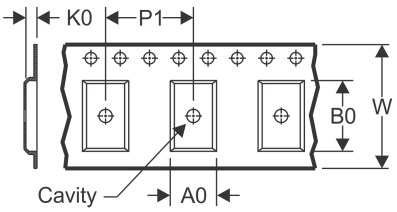
(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

TAPE AND REEL INFORMATION

REEL DIMENSIONS

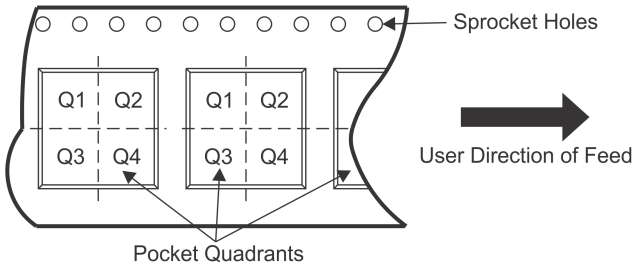


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

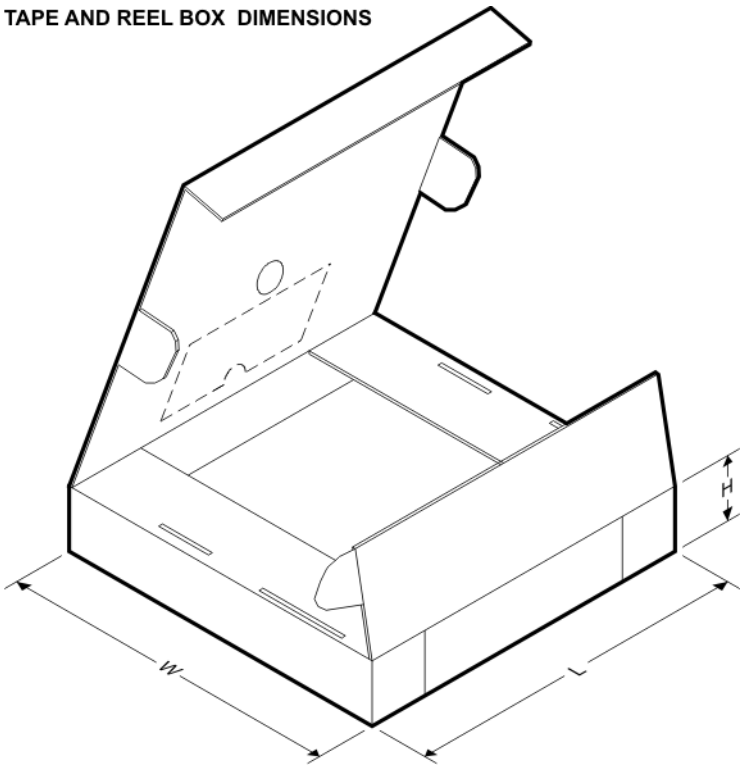
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS3711DDCR	SOT-23-THIN	DDC	6	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TPS3711DDCT	SOT-23-THIN	DDC	6	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

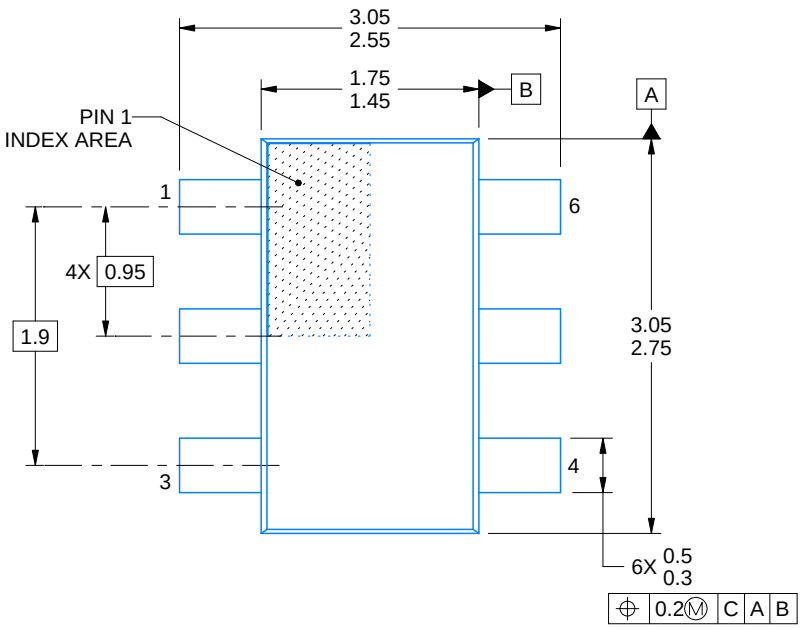
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS3711DDCR	SOT-23-THIN	DDC	6	3000	195.0	200.0	45.0
TPS3711DDCT	SOT-23-THIN	DDC	6	250	195.0	200.0	45.0

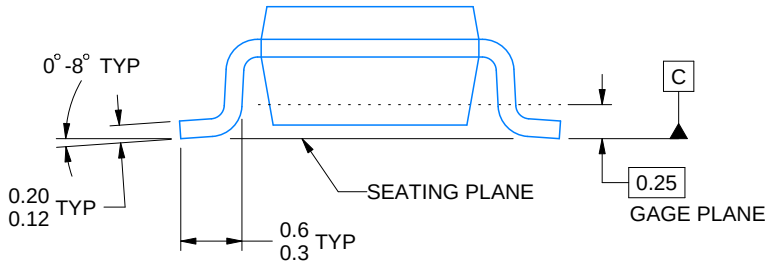
PACKAGE OUTLINE

DDC0006A

SOT - 1.1 max height

SOT





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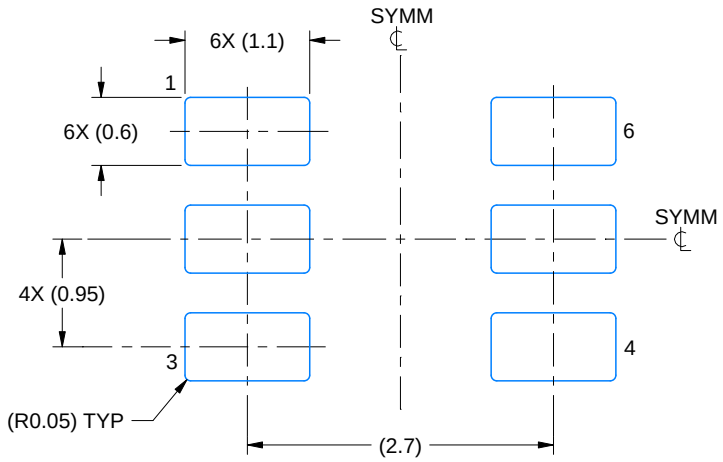
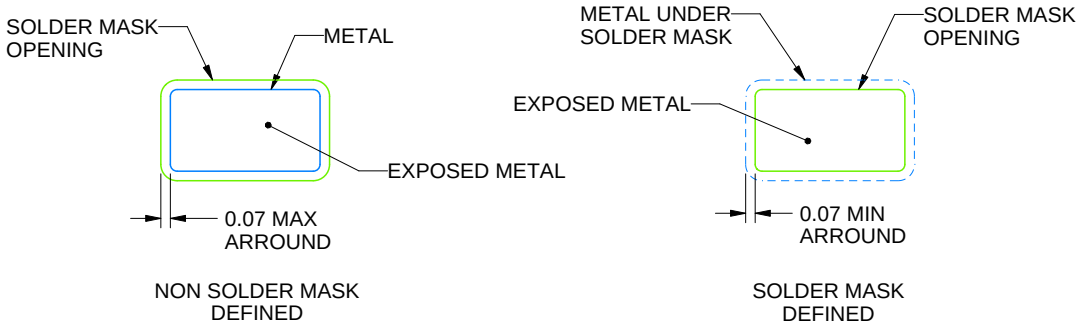
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Reference JEDEC MO-193.



LAND PATTERN EXAMPLE
EXPLOSED METAL SHOWN
SCALE:15X



SOLDERMASK DETAILS

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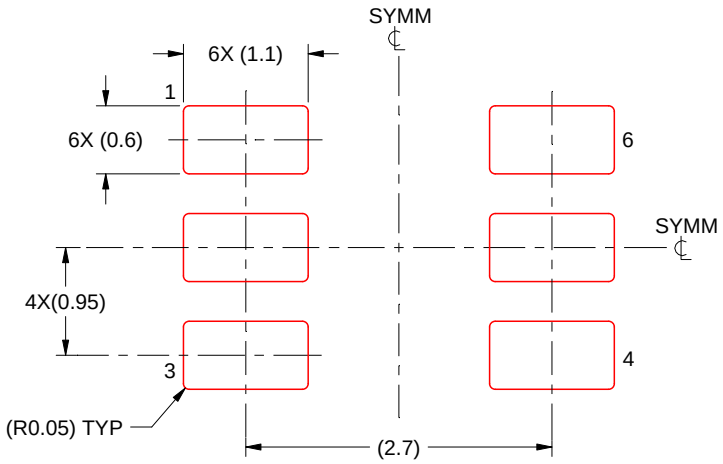
NOTES: (continued)

4. Publication IPC-7351 may have alternate designs.
5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOLDER PASTE EXAMPLE
BASED ON 0.125 THICK STENCIL
SCALE:15X

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
7. Board assembly site may have different recommendations for stencil design.

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